

# ARBITRATE-AND-MOVE PRIMITIVES FOR HIGH THROUGHPUT ON-CHIP INTERCONNECTION NETWORKS\*

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## ABSTRACT

An  $n$ -leaf pipelined balanced binary tree is used for arbitration of order and movement of data from  $n$  input ports to one output port. A novel *arbitrate-and-move* primitive circuit for every node of the tree, which is based on a concept of *reduced synchrony* that benefits from attractive features of both asynchronous and synchronous designs, is presented. The design objective of the pipelined binary tree is to provide a key building block in a high-throughput mesh-of-trees interconnection network for Explicit Multi Threading (XMT) architecture, a recently introduced parallel computation framework. The proposed reduced synchrony circuit was compared with asynchronous and synchronous designs of arbitrate-and-move primitives. Simulations with  $0.18\mu\text{m}$  technology show that compared to an asynchronous design, the proposed reduced synchrony implementation achieves a higher throughput, up to 2 Giga-Requests per second on an 8-leaf binary tree. Our circuit also consumes less power than the synchronous design, and requires less silicon area than both the synchronous and asynchronous designs.

## 1. INTRODUCTION

In spite of the continuing fast increase in the number of transistors that can fit on a single chip (the *Billion-transistor chip era*), the speed of computer processors has entered a state of diminishing returns in offering growth beyond clock speed. This is happening as scientific and commercial workloads keep expanding at a fast rate.

The outreach of parallel computing has not met some expectations because of parallel computer systems' programmability shortcomings. The Parallel Random Access Model (PRAM) has been developed by numerous algorithm researchers during the 1980s and 1990s, as the key to coping with the programmability challenge. However, it had not been possible to build parallel machines using multi-chip multiprocessors, the only multiprocessors buildable in the 1990s, to support PRAM. The Explicit Multi-Threading (XMT) project ([16], [8]) at the University of Maryland is based on the insight that it is becoming possible to build a parallel processor, which can be programmed with a PRAM-like language on a single chip as silicon capacity continues to increase.

The performance of XMT depends on communication between processor clusters and memory modules. A memory subsystem has been proposed at the logic and physical level for the XMT architecture ([9]). It uses a *mesh of pipelined balanced binary trees* to implement a crossbar type interconnection network. At each node of each binary tree, data buffers and arbiter circuits will be used. Due to the large number of such arbitration circuits in the system, their throughput, speed and

power consumption will have significant impact on the memory subsystem's performance. The goal of this paper is to describe a fast and power-efficient implementation of the arbitration circuit at circuitry level. Before elaborating on the difference between our approach and the state-of-the-art arbiters, we mention that our arbiter circuit is not limited to XMT architecture. It can be used to help implementing other on-chip high throughput and low power interconnection networks.

Research on state-of-the-art arbiter circuits were motivated by applications, where the performance of crossbar switches and bus-based networks is directly related to the latency of arbitration (e.g. [4], [10], [13] and [18]). These works focused on  $n$ -to-1 arbitration, where a *request vector* ('1' means a request) of  $n$ -bits is given, and an  $n$ -bit *grant vector* is generated with only one granting signal among  $n$  bits.

In cases of crossbar network applications ([4], [13] and [18]), the grant vector is used to configure the switches to connect input ports to output ports. Communication (moving data from input to output port) follows switch configuration. In case of bus network applications ([10]), the grant signal decides which source owns the bus in the next cycle to transmit data. In all of the above approaches, arbitration among  $n$  elements precedes the data movement. In contrast, we combine arbitration and data movement. We move the data one level up in the binary tree, towards the root (output port), as soon as we arbitrate between two neighboring requests. Arbitration and movement repeat at each stage until the data reaches the root. Each stage of the pipeline has the latency of a single (2-to-1) *arbitrate-and-move* primitive, which promises great increase in throughput.

The latency of traditional arbiter circuits is measured from the instant the request vector is modified to the instant when the grant vector is updated.  $N$ -bit arbiter circuits in each of [10], [13] and [18], were built in a tree structure, using 2-input or 4-input primitives. Critical delay path, which influences latency directly, extends from leaves to the root ([10]) and in some cases, from leaves to the root and then back to leaves ([13], [18]). [4] proposes a two-level design to reduce latency, which may work well for high number of inputs ( $n > 2$ ), however for 2-input arbiters a single level of logic gates would be sufficient. The arbiters, proposed in [4] and [13] are based on *priority encoder* circuits. Priority encoders are simpler and faster, compared to circuits in [10] and [18]; however, they require support of extra circuits to provide fairness among requesting inputs. The above  $n$ -to-one arbiter circuits are not as useful for our purpose of implementing the pipelined binary tree. Their underlying complex algorithms are optimized for  $n > 2$  inputs implying higher latencies for the primitives, and therefore lower performance when pipelining is used.

The rest of the paper is organized as follows. We state the problem and describe our approach in Section 2. The arbiter circuits are presented in Section 3 and simulation results are reported in Section 4. Section 5 concludes the paper with a discussion of our current and ongoing work.

For an extended version of this paper, see the XMT home page <http://www.umiacs.umd.edu/users/vishkin/XMT/>.

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These circuits are not connected to a global clock directly. An external *fast-clock* signal is connected to the root circuit only. The children receive clock pulses from their parents, only when needed. There is synchrony between a parent and its immediate children, but there is no global synchrony as in a fully synchronous circuit. Hence, we use the term, *reduced synchrony*.

We initially designed such a circuit using static CMOS gates. Further observations suggested that a design with dynamic gates could yield better performance in terms of speed and power.

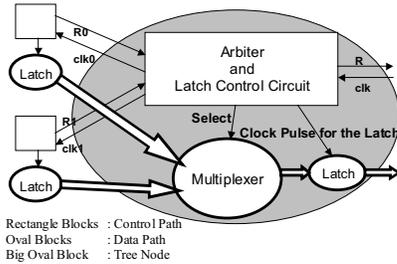


Figure 2: RS-Static Arbitrate-and-Move Node

### 3.3.1. Static Gate Implementation (RS-Static)

The node is connected to its children as shown in Figure 2. It operates as follows:

- If no requests come from the children, no request signal is sent to the parent, and both children receive a clock pulse.
  - If a single request comes from a child, it passes, and the clock pulse is passed to both of the children.
  - If two requests come, one passes first and then the other.
- Only the child with passing request receives a clock pulse.

The implementation can be seen in Figure 3. The critical delay, which determines the clock period of this circuit, is the amount of time between the generation of the clock pulse at the parent node and the update of the request signal at the child node. The former consists of a 2-input *nand* gate and two *inverters*, and the latter consists of a D-latch, implemented as a *transmission gate* and 2 *inverters*. Since some of these gates have a fan-out of 3 or 4, we cannot achieve the delay of a gate of the reference ring oscillator as described for the asynchronous circuit in Section 3.1.

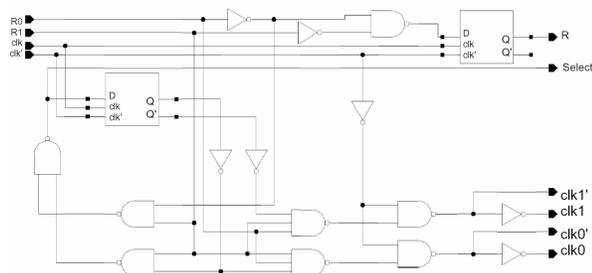


Figure 3: Schematic of RS-Static Arbitrate-and-Move Node Circuit

### 3.3.2. Dynamic Gate Implementation (RS-Dynamic)

The high-level node structure is the same as in Figure 2 except that the clock signals to the children are unified to *clk-out* signal. Dynamic logic gates from True Single Phase Clocking (TSPC) family ([17]) are used for this implementation (Figure 4). Each gate executes a simple logic function and latches the result for one clock period. This allows the parent node to modify signals of its children. The high-level algorithm is as follows:

- If no requests come from the children, no request signal is sent to the parent, and both children receive a clock pulse.
- If a single request comes from a child, it passes, and the clock pulse is passed to both of the children.

- If two requests come, the one from Child 0 passes, then the parent kills that request. The clock signal is not passed to the children. (At the next cycle, the request at Child 1 remains but Child 0 does not generate a new request)

If both children send a request at a given cycle, the request of child 1 passes as the *only remaining request* at the next cycle. Therefore, arbitration is fair, despite of the built-in priority of Child 0. (Figure 5)

The critical delay path is similar to that of RS-Static, however fewer gates are used: Clock pulse is generated through two half gates (dynamic gates) and a buffer. Request is generated through one dynamic gate.

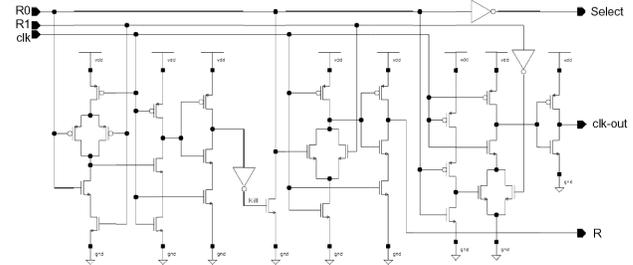


Figure 4: Schematic of RS-Dynamic Arbitrate-and-Move Node Circuit

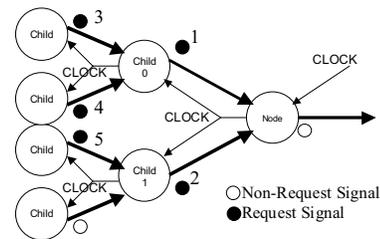


Figure 5: As the Node receives continuous clock pulses, requests pass in order 1, 2, 3, 5, and 4 through the Node.

## 4. SIMULATION RESULTS

In this section, we describe our simulation setup for the study of the proposed circuits and report our preliminary results. We used Cadence tools (SPECTRE simulator) for all simulations and delay and power measurements. SPICE parameters for 0.18 $\mu$ m technology are obtained from [7]. A 1.8V source is used.

We simulated a single node of the asynchronous arbitrate-and-move circuit with minimum sized transistors, and an 8-leaf-7-node binary tree for both RS-Static and RS-Dynamic circuits. RS-Dynamic (Figure 4) is optimized using Cadence Analog Circuit Optimizer. RS-Static (Figure 3) is optimized only manually, because our preliminary estimations did not suggest that it is promising in outperforming RS-Dynamic. For the same reason and due to space constraint, we will not elaborate on the synchronous arbiter circuit.

As reference, we built a 7-stage ring oscillator using inverters with a  $W_p/W_n$  ratio of 13/6 at the same technology. Each inverter showed a 49ps propagation delay (50% of the input to 50% of the output) of full voltage swing) and a 73ps rise time (10% to 90% of full voltage swing). We thus targeted a rise time of 80ps for the clock pulse when we simulate the arbitrate-and-move circuits.

Table 1 reports the performance, measured by speed, area, and power, of different circuit implementations for a test design explained above. The root node of RS circuits is connected to a clock generator with various clock periods for our simulation. Speed is measured as the fastest clock (or equivalently, the shortest clock period) that the circuit can keep pace with for the RS circuits, and the time interval between the request and done signals for the asynchronous circuit. Area is measured as the

total transistor area (width x length) and the number of transistors required per node circuit. Power consumption of each circuit is measured as the average power consumed by the global power source on the same request pattern.

The shortest clock period for the RS-static circuit to work correctly is 800ps, while the RS-dynamic circuit operates correctly at a clock period as fast as 500ps. (The critical delay path of RS-dynamic circuit is even shorter, but we observed that the clock signal may not be able to complete full voltage swing occasionally). The asynchronous circuit completes one arbitration cycle in 2.5ns. No other requests can be processed during this time. The synchronous circuit was not measured as it is clearly dominated by the proposed reduced synchrony circuits.

Node Circuit	Speed	Transistor Area	No. of transistors	Power consumption
RS-Dynamic	500 ps	28.4 $\mu\text{m}^2$	34	16.4 mW
RS-Static	800 ps	38.4 $\mu\text{m}^2$	72	8.8 mW
Asynchronous	2.5 ns	$\geq 35\mu\text{m}^2$	188	N/A
Synchronous	N/A	N/A	$\sim 90$ (est.)	N/A

**Table 1: Performance comparison of the four different arbitrate-and-move circuit implementations.**

## 5. DISCUSSION

All circuits described above contain state holding elements. In the asynchronous one, these elements may go into a metastable state as the node receives two request signals concurrently. Extra circuitry, which reduces the speed of the circuit, is required to prevent this from happening. In the synchronous circuit, the inputs of the state holding elements have to be ready within the same clock period and a stalling condition at the root may delay inputs at the leaf circuits. Both of these properties cause degradation of throughput in an overall system.

Power consumption is proportional to switching frequency, and load capacitance. Although we have not implemented the synchronous arbiter circuit, we expect that it will consume more power because of the clock network (clock generator and the clock tree) and the fact that all clocked gates will operate at the highest frequency. On the other hand, in the RS circuits, a node sends clock pulse to its children only when it is ready for a new request. Therefore the inner nodes will not always receive the fastest clock signal. In the case when the interconnect tree is fully loaded with requests, there will be only one single branch driven by the clock, from the root towards the leaves, and the remaining part of the tree will not consume any power. Table 1 shows that RS-Static is slower but consumes less power compared to the RS-Dynamic circuit. A more realistic result could be observed if we would let the input pattern change over time, as will be the case in the XMT implementation. Future research on XMT will need to examine load characteristic on the interconnection network.

In sum, we conclude that the dynamic gate implementation of reduced synchrony circuit best suits our design objectives, while the static gate implementation is also useful when compared with traditional asynchronous or synchronous designs. As for future work, we first mention that opportunities for less power consumption still exist; for example, an extra circuit to avoid unnecessary precharging of the dynamic gates. Also, note that the total silicon area of the circuit should also include diffusion areas, overhead areas, pins and metal wires among others. At the current stage of this project, we are unable to determine these values without the silicon layout. However, we expect that the circuit with larger transistor area will have larger total silicon area. Finally, we mention that, when we compare area and power consumption with the synchronous design, any clock driver required by the synchronous arbiter must also be included for a fair comparison, because the clock circuit of the reduced synchrony arbiter circuits is built into the nodes.

## 6. CONCLUSION

We reviewed *asynchronous* and *synchronous* arbiter circuits, and proposed *reduced synchrony* arbiter circuits. Simulations show that a binary tree shaped arbitration network can be implemented by our circuit in a fast (2 Giga-Requests/second at 0.18 $\mu\text{m}$ ) and power effective way. The modularity of our approach, where the input characteristics of the primitive circuit match its output characteristics, provides design scalability.

We believe that future work on XMT could greatly benefit from the throughput provided by this circuit. Furthermore the proposed an arbitration scheme may prove useful in areas where performance is held back due to high latency of present arbitration and scheduling schemes, such as issue-logic of wide-issue superscalar processors [3].

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