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Robust Output Feedback Control for Discrete - Time Nonlinear Systems

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Abstract

In this paper we present a new approach to the solution of the output feedback robust control problem. We employ the recently developed concept of *information state* for output feedback dynamic games, and obtain necessary and sufficient conditions for the solution to the robust control problem expressed in terms of the information state. The resulting controller is an information state feedback controller, and is intrinsically *infinite dimensional*. Stability results are obtained using the theory of dissipative systems, and indeed, our results are expressed in terms of *dissipation inequalities*.

Key words: Output feedback robust control, nonlinear control systems, information state, stability, bounded real lemma, dissipative systems.

1 Introduction

The modern theory of robust (or H_∞) control for linear systems originated in the work of Zames [28], which employed frequency domain methods (see also Zames and Francis [8], [29]). After the publication of this work, there was an explosion of research activity which led to a rather complete and satisfying body of theory, see Doyle *et al* [6]. In fact, the successful development of this theory is, to a large extent, due to the use of time domain methods. In addition, significant advances in the theory depended on ideas

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from elsewhere; in particular, extensive use has been made of results concerning dynamic games, Riccati equations, the bounded real lemma (e.g., Basar and Bernhard [4], Doyle *et al* [6], Limebeer *et al* [16], Petersen *et al* [18], [19]), and risk-sensitive stochastic optimal control (e.g., Glover and Doyle [7], Whittle [25]). The solution to the output feedback robust control problem has the structure of an observer and a controller, and involves filter and control type Riccati equations.

The time domain formulation of the robust control problem has a natural generalization to nonlinear systems, since the H_∞ norm inequality $\|\Sigma\|_{H_\infty} < \gamma$ has an interpretation which in no way depends on linearity (of course, use of the term “norm” may not be appropriate for nonlinear systems). This inequality is related to the L_2 gain of the system and the bounded real lemma. The robust control problem is to find a stabilizing controller which achieves this H_∞ norm bound, and can be viewed as a dynamic game problem, with nature acting as a malicious opponent. A general and powerful framework for dealing with L_2 gains for nonlinear systems is Willems’ theory of dissipative systems [27]. Using this framework, one can write down a nonlinear version of the bounded real lemma, which is expressed in terms of a dynamic programming inequality or a partial differential inequality, known as the dissipation inequality (see, e.g., Hill and Moylan [10]), which reduces to a Riccati inequality or equation in the linear context. Therefore, it is not surprising that in papers dealing with nonlinear robust control, one sees dissipation inequalities and equations and dynamic game formulations (e.g., Ball *et al* [2], [3], Isidori and Astolfi [12], van der Schaft [21], [22], [23]).

An examination of the references cited above reveals that the state feedback robust control problem for nonlinear systems is reasonably well understood: one obtains the controller by solving the dissipation-type inequality or equation which results from the dynamic game formulation (actually, controller synthesis remains a major difficulty for continuous-time systems, but the conceptual framework is in place). The output feedback case is not nearly so well developed, and no general framework for solving it is available in the literature. By analogy with the linear case, one expects the solution to involve a filter or observer in addition to a dissipation inequality/equation for determining the control. Several authors have proceeded by postulating a filter structure and solving an augmented game problem, [3], [12], [23]. These results yield sufficient conditions, which are in general not necessary conditions; that is, an output feedback problem may be solvable, but not necessarily by the means that have thus far been suggested.

In this paper we present a new approach to the solution of the output feedback robust control problem for nonlinear systems. Our approach yields conditions which are both necessary and sufficient. The framework we present incorporates a *separation principle*, which in essence permits the replacement of the original output feedback problem by an equivalent one with full information, albeit infinite dimensional. (The continuous-time problem is also solvable using our approach, at least in principle [14]. The relevant continuous-time dissipation inequality can readily be written down; however, rigorous results will require considerable effort. This will be the subject of future papers. Here, we discuss only discrete-time systems.)

Our approach to this problem was motivated by ideas from stochastic control and large deviations theory. In our earlier paper [13], we explored the connection between a partially observed risk-sensitive stochastic control problem and a partially observed dynamic game, and we introduced the use of an *information state* for solving such games. The information state for this game was obtained as an asymptotic limit of the information state for the risk-sensitive stochastic control problem. Historically, the information state we employ is related to the “past stress” used by Whittle [25] in solving the risk-sensitive problem for linear systems (see also [26]), and can be thought of as a modified conditional density or minimum energy estimator (c.f. Hijab [9], Mortensen [17]). Basar and Bernhard [4] also use the past stress for solving game problems for linear systems. The framework developed in this paper to solve the output feedback robust control problem involves a dynamic game formulation, and the use of the (infinite dimensional) information state dynamical system constitutes the above-mentioned separation principle. This idea of separation, using, say, the conditional density, is well known in stochastic control theory, see Kumar and Varaiya [15]. Our results imply that if the robust control problem is at all solvable by an output feedback controller, then it is solvable by an *information state feedback controller*.

The information state feedback controller we obtain has an observer/controller structure. The “observer” is the dynamical system for the information state $p_k(x)$:

$$p_k = F(p_{k-1}, u_{k-1}, y_k)$$

(the notation is introduced in §2). The “controller”

$$u_k = \bar{u}^*(p_k)$$

is determined by a dynamic programming inequality,

$$W(p) \geq \inf_{u \in U} \sup_{y \in \mathbb{R}^p} \{ W(F(p, u, y)) \};$$

and the value function $W(p)$ solving it is a function of the information state. This dynamic programming inequality is an infinite dimensional relation defined for an infinite dimensional control problem, namely that of controlling the information state. Our solution is therefore an *infinite dimensional dynamic compensator*, Figure 1. In a sense, the solution is “doubly infinite dimensional”.

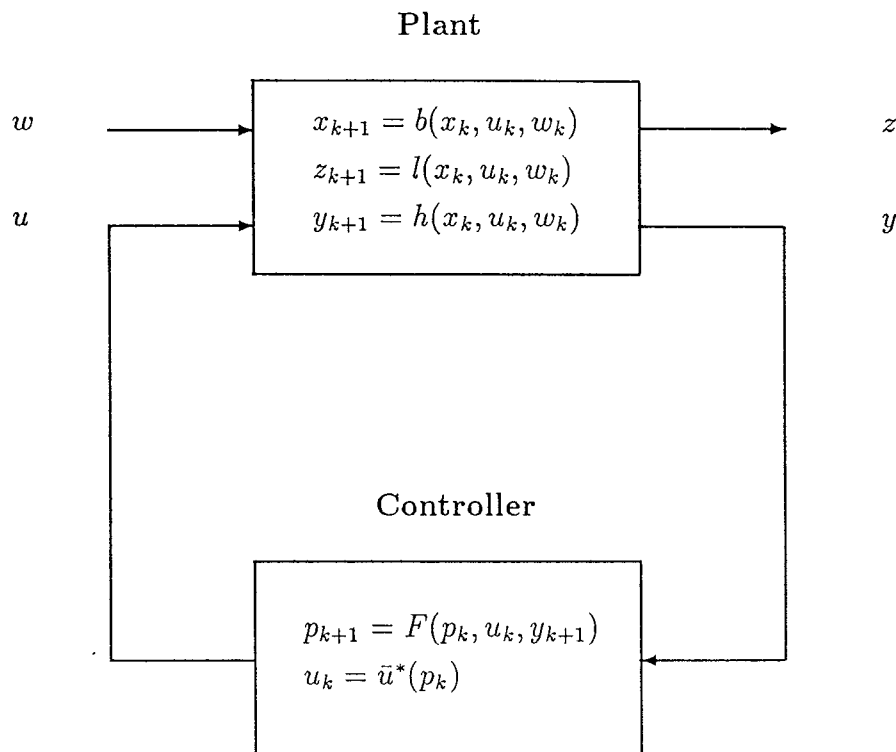


Figure 1

While there is a separation principle, the task of “estimation” is not isolated from that of “control”. The information state carries observable information that is relevant to the control objective, and need not necessarily accurately estimate the state of the system being controlled. The control objective is taken into consideration and so the resulting state estimate is suboptimal, but nonetheless more suitable to achieving the control objective, relative to an observer designed with the exclusive aim of state estimation. Thus the information state represents the optimal trade-off between estimation and control for the robust control problem.

We begin in §2 by formulating the problem to be solved. Then in §3 we consider the state feedback problem; it is hoped that our treatment of this problem will clarify certain aspects of our solution to the output feedback problem, which is presented in §4. Note, however, that the solution to the state feedback problem is *not* used to solve the output feedback problem. Our results are obtained in a rather general context, and as a consequence the use of extended-real valued functions is necessary; of course, if one imposes various regularity and non-degeneracy conditions, this can be avoided. We remark that while the key ideas for our solution were obtained from stochastic control theory, this paper makes no explicit use of that theory, and is in fact self-contained and purely deterministic.

2 Problem Formulation

We consider discrete-time nonlinear systems (plants) Σ described by the state space equations of the general form

$$(1) \quad \begin{cases} x_{k+1} = b(x_k, u_k, w_k), \\ z_{k+1} = l(x_k, u_k, w_k), \\ y_{k+1} = h(x_k, u_k, w_k). \end{cases}$$

Here, $x_k \in \mathbf{R}^n$ denotes the state of the system, and is not in general directly measurable; instead an output quantity $y_k \in \mathbf{R}^p$ is observed. The additional output quantity $z_k \in \mathbf{R}^q$ is a performance measure, depending on the particular problem at hand. The control input is $u_k \in U \subset \mathbf{R}^m$, and $w_k \in \mathbf{R}^r$ is a disturbance input. For instance, w could be due to modelling errors, sensor noise, etc. The system behavior is determined by the functions $b : \mathbf{R}^n \times \mathbf{R}^m \times \mathbf{R}^r \rightarrow \mathbf{R}^n$, $l : \mathbf{R}^n \times \mathbf{R}^m \times \mathbf{R}^r \rightarrow \mathbf{R}^q$, $h : \mathbf{R}^n \times \mathbf{R}^m \times \mathbf{R}^r \rightarrow \mathbf{R}^p$. It is assumed that the origin is an equilibrium for the system (1): $b(0, 0, 0) = 0$, $l(0, 0, 0) = 0$, and $h(0, 0, 0) = 0$.

The *output feedback robust control problem* is: given $\gamma > 0$, find a controller $u = u(y(\cdot))$, responsive only to the observed output y , such that the resulting closed loop system Σ^u achieves the following two goals;

1. Σ^u is asymptotically stable when no disturbances are present, and
2. Σ^u is *finite gain*, i.e., for each initial condition $x_0 \in \mathbf{R}^n$ the input-output map $\Sigma_{x_0}^u$ relating w to z is finite gain, which means that there exists a finite quantity $\beta^u(x_0)$ such that

$$(2) \quad \begin{cases} \sum_{i=0}^{k-1} |z_{i+1}|^2 \leq \gamma^2 \sum_{i=0}^{k-1} |w_i|^2 + \beta^u(x_0) \\ \text{for all } w \in \ell_2([0, k-1], \mathbf{R}^r) \text{ and all } k \geq 0. \end{cases}$$

Since $x_0 = 0$ is an equilibrium, we also require that $\beta^u(0) = 0$.

Of course, β will also depend on γ .

Note that we have specified the robust control problem in terms of the family of initialized input-output maps $\{\Sigma_{x_0}^u\}_{x_0 \in \mathbf{R}^n}$, whereas the conventional problem statement for linear systems refers only to the single map Σ_0^u . This is often expressed in terms of the H_∞ norm of Σ_0^u :

$$(3) \quad \|\Sigma_0^u\|_{H_\infty} \triangleq \sup_{w \in \ell_2([0, \infty), \mathbf{R}^r), w \neq 0} \frac{\|z\|_{\ell_2([1, \infty), \mathbf{R}^q)}}{\|w\|_{\ell_2([0, \infty), \mathbf{R}^r)}}.$$

For linear systems, the linear structure means that the solvability of the robust control problem is equivalent to the solvability of a pair of Riccati difference equations (and a

coupling condition), under certain assumptions, and so implicitly all the maps $\Sigma_{x_0}^u$ are considered. For nonlinear systems, our formulation seems natural and appropriate (see [10], [24]), since otherwise if we were to follow the linear systems formulation, one would need assumptions relating non-zero initial states x_0 to the equilibrium state 0 (such as reachability). The formulation adopted here has also been used recently by van der Schaft [23]. A solution $u = u^*$ to this problem yields

$$(4) \quad \|\Sigma_0^u\|_{H_\infty} \leq \gamma,$$

as is the case for linear systems. It is also apparent that in place of the ℓ_2 norm used in the definition of the finite gain property, one could substitute any other ℓ_q norm [24], or indeed, any other suitable function, and the corresponding theory would develop analogously.

3 The State Feedback Case

In this section we consider the special case where complete state information is available, i.e., where $h(x, u, w) \equiv x$. It is *not* assumed that the disturbance is measured. For an alternative presentation of the state feedback problem, see [2].

3.1 Problem

The system Σ is now described by

$$(5) \quad \begin{cases} x_{k+1} = b(x_k, u_k, w_k), \\ z_{k+1} = l(x_k, u_k, w_k), \end{cases}$$

where $u \in \mathcal{S}$ is a state feedback controller, i.e., those controllers for which $u_k = \bar{u}(x_k)$, where $\bar{u} : \mathbf{R}^n \rightarrow U$.

The *state feedback robust control problem* is: given $\gamma > 0$, find a feedback controller $u \in \mathcal{S}$ such that the resulting closed loop system Σ^u achieves the following two goals;

1. Σ^u is asymptotically stable when no disturbances are present, and
2. Σ^u is *finite gain*, i.e., for each initial condition $x_0 \in \mathbf{R}^n$ the corresponding input-output map $\Sigma_{x_0}^u$ relating w to z is finite gain, which means that there exists a finite quantity $\beta^u(x_0)$, with $\beta^u(0) = 0$, such that

$$(6) \quad \begin{cases} \sum_{i=0}^{k-1} |z_{i+1}|^2 \leq \gamma^2 \sum_{i=0}^{k-1} |w_i|^2 + \beta^u(x_0) \\ \text{for all } w \in \ell_2([0, k-1], \mathbf{R}^r), \text{ and all } k \geq 0. \end{cases}$$

Before attacking the full problem, we consider the finite time problem, where stability is not an issue.

3.2 Finite Time Case

Let $\mathcal{S}_{k,l}$ denote the set of controllers u defined on the time interval $[k, l]$ such that for each $j \in [k, l]$ there exists a function $\bar{u}_j : \mathbf{R}^{(j-k+1)n} \rightarrow U$ such that $u_j = \bar{u}_j(x_{k,j})$.

The *finite time state feedback robust control problem* is: given $\gamma > 0$ and a finite time interval $[0, k]$, find a feedback controller $u \in \mathcal{S}_{0,k}$ such that the resulting closed loop system Σ^u achieves the following goal;

Σ^u is *finite gain*, i.e., for each initial condition $x_0 \in \mathbf{R}^n$ the corresponding input-output map $\Sigma_{x_0}^u$ relating w to z is finite gain, which means that there exists a finite quantity $\beta_k^u(x_0)$, with $\beta_k^u(0) = 0$, such that

$$(7) \quad \begin{cases} \sum_{i=0}^{k-1} |z_{i+1}|^2 \leq \gamma^2 \sum_{i=0}^{k-1} |w_i|^2 + \beta_k^u(x_0) \\ \text{for all } w \in \ell_2([0, k-1], \mathbf{R}^r). \end{cases}$$

3.2.1 Dynamic Game

For $u \in \mathcal{S}_{0,k-1}$ and $x_0 \in \mathbf{R}^n$ we define the functional $J_{x_0,k}(u)$ for (5) by

$$(8) \quad J_{x_0,k}(u) = \sup_{w \in \ell_2([0,k-1], \mathbf{R}^r)} \left\{ \sum_{i=0}^{k-1} |z_{i+1}|^2 - \gamma^2 |w_i|^2 : x(0) = x_0 \right\}.$$

Clearly,

$$J_{x_0,k}(u) \geq 0,$$

and the finite gain property of $\Sigma_{x_0}^u$ can be expressed in terms of J as follows:

Lemma 3.1 $\Sigma_{x_0}^u$ is finite gain on $[0, k]$ if and only if there exists a finite quantity $\beta_k^u(x_0)$ such that

$$(9) \quad J_{x_0,j}(u) \leq \beta_k^u(x_0), \quad j \in [0, k],$$

and $\beta_k^u(0) = 0$.

The *state feedback* dynamic game is to find a control $u^* \in \mathcal{S}_{0,k-1}$ which minimizes each functional $J_{x_0,k}$, $x_0 \in \mathbf{R}^n$. This will yield a solution to the finite time state feedback robust control problem.

3.2.2 Solution to the Finite Time State Feedback Robust Control Problem

The dynamic game can be solved using dynamic programming [4]. The idea is to use the value function

$$(10) \quad V_j(x) = \inf_{u \in \mathcal{S}_{0,j-1}} \sup_{w \in \ell_2([0,j-1], \mathbf{R}^r)} \left\{ \sum_{i=0}^{j-1} |z_{i+1}|^2 - \gamma^2 |w_i|^2 : x(0) = x \right\},$$

and corresponding dynamic programming equation

$$(11) \quad \begin{cases} V_j(x) = \inf_{u \in U} \sup_{w \in \mathbf{R}^r} \{V_{j-1}(b(x, u, w)) + |l(x, u, w)|^2 - \gamma^2 |w|^2\} \\ V_0(x) = 0. \end{cases}$$

Theorem 3.2 (Necessity) *Assume that $u^s \in \mathcal{S}_{0,k-1}$ solves the finite time state feedback robust control problem. Then there exists a solution V to the dynamic programming equation (11) such that $V_j(0) = 0$ and $V_j(x) \geq 0$, $j \in [0, k]$.*

PROOF. For $x \in \mathbf{R}^n$, $j \in [0, k]$, define $V_j(x)$ by the formula (10), i.e.,

$$V_j(x) = \inf_{u \in \mathcal{S}_{0,j-1}} J_{x,j}(u).$$

Then we have

$$0 \leq V_j(x) \leq \beta_k^{u^*}(x), \quad j \in [0, k].$$

Thus V is finite, and since $\beta_k^{u^*}(0) = 0$, $V_j(0) = 0$. By dynamic programming (e.g. [4]), V satisfies the equation (11). \square

Theorem 3.3 (Sufficiency) *Assume there exists a solution V to the dynamic programming equation (11) such that $V_j(0) = 0$ and $V_j(x) \geq 0$, $j \in [0, k]$. Let $u^* \in \mathcal{S}_{0,k-1}$ be a policy such that $u_j^* = \bar{u}_{k-j}^*(x_j)$, where $\bar{u}_j^*(x)$ achieves the minimum in (11); $j = 0, \dots, k-1$. Then u^* solves the finite time state feedback robust control problem.*

PROOF. Standard dynamic programming arguments imply that

$$V_k(x) = J_{x,k}(u^*) = \inf_{u \in \mathcal{S}_{0,k-1}} J_{x,k}(u),$$

and so u^* is an optimal policy for the game. In particular,

$$J_{x_0,k}(u^*) = V_k(x_0),$$

for all $x_0 \in \mathbf{R}^n$. Then applying Lemma 3.1, we see that (9) is satisfied with $u = u^*$ and $\beta_k^u(x) \triangleq V_k(x)$. \square

3.3 Infinite Time Case

We wish to solve the infinite time problem by passing to the limit

$$(12) \quad \lim_{k \rightarrow \infty} V_k(x) = V(x),$$

where $V_k(x)$ is defined by (10), to obtain a stationary version of the dynamic programming equation (11), viz.,

$$(13) \quad V(x) = \inf_{u \in U} \sup_{w \in \mathbf{R}^r} \left\{ V(b(x, u, w)) + |l(x, u, w)|^2 - \gamma^2 |w|^2 \right\}.$$

In many respects, this procedure is best understood in terms of the Bounded Real Lemma [1], [19]. For instance, the finite gain property is captured in terms of a *dissipation inequality* [27] (or partial differential inequality in continuous-time). Also, stability results are readily deduced [10], [27].

3.3.1 Bounded Real Lemma

We will say that Σ^u is *finite gain dissipative* if there exists a function (called a *storage function* [27]) $V(x)$ such that $V(x) \geq 0$, $V(0) = 0$, and satisfies the dissipation inequality

$$(14) \quad V(x) \geq \sup_{w \in \mathbf{R}^r} \left\{ V(b(x, \bar{u}(x), w)) - \gamma^2 |w|^2 + |l(x, \bar{u}(x), w)|^2 \right\}.$$

Theorem 3.4 (Bounded Real Lemma.) *Let $u \in \mathcal{S}$. The system Σ^u is finite gain if and only if it is finite gain dissipative.*

PROOF. If Σ^u is finite gain dissipative, then (14) implies

$$(15) \quad V(x_k) + \sum_{i=0}^{k-1} |z_{i+1}|^2 \leq \gamma^2 \sum_{i=0}^{k-1} |w_i|^2 + V(x_0),$$

for all $k > 0$ and all $w \in \ell_2([0, k-1], \mathbf{R}^r)$. Using the non-negativity of V this inequality yields (6) with $\beta^u(x_0) = V(x_0)$. Therefore Σ^u is finite gain.

Conversely, assume that Σ^u is finite gain. Then we have

$$0 \leq J_{x_0, k}(u) \leq \beta^u(x_0)$$

for all $k \geq 0$, where β does not depend on k . Write $V_k(x) = J_{x, k}(u)$, so that

$$0 \leq V_k(x) \leq \beta^u(x), \quad \text{for all } x \in \mathbf{R}^n, \quad k \geq 0.$$

Now $V_k(x)$ enjoys the monotonicity property

$$V_k(x) \leq V_{k+1}(x),$$

and so the limit

$$(16) \quad V_a(x) = \lim_{k \rightarrow \infty} V_k(x)$$

exists and is finite. Dynamic programming implies that V_a solves the dissipation inequality (14):

$$V_a(x) \geq \sup_{w \in \mathbf{R}^r} \left\{ V_a(b(x, \bar{u}(x), w)) + |l(x, \bar{u}(x), w)|^2 - \gamma^2 |w|^2 \right\}.$$

Also, it is clear that $V_a(0) = 0$ and $V_a(x) \geq 0$. Therefore, V_a is a storage function, and so Σ^u is finite gain dissipative. \square

Remark 3.5 If Σ^u is *reachable* (from 0), that is, if for any $x \in \mathbf{R}^n$ there exists $k_1 < 0$ and $w \in \ell_2([k_1, -1], \mathbf{R}^r)$ such that $x_{k_1} = 0$ and $x_0 = x$, then the finite gain property for the *single* input-output map Σ_0^u implies the finite gain property for *all* maps $\Sigma_{x_0}^u$. To see this, select $k > 0$, $w \in \ell_2([0, k-1], \mathbf{R}^r)$, and $x \in \mathbf{R}^n$. There exists $k_1 < 0$ and $\bar{w} \in \ell_2([k_1, -1], \mathbf{R}^r)$ such that $x_{k_1} = 0$ and $x_0 = x$. Define

$$\tilde{w} = \begin{cases} \bar{w} & \text{on } [k_1, -1], \\ w & \text{on } [0, k-1]. \end{cases}$$

Since Σ^u is time-invariant, (6) implies

$$\sum_{i=k_1}^{k-1} |z_{i+1}|^2 - \gamma^2 |\tilde{w}_i|^2 \leq 0,$$

and hence

$$\sum_{i=0}^{k-1} |z_{i+1}|^2 - \gamma^2 |w_i|^2 \leq \phi(x),$$

for some finite quantity ϕ depending only on x . Therefore $V_a(x)$, defined by (16) is bounded above by $\beta^u(x) \triangleq \phi(x)$. Note that $\phi(0) = 0$. \square

Remark 3.6 The function V_a defined by (16) is known as the *available storage* [27]. If Σ^u is finite gain dissipative with storage function V , then $V_a \leq V$, and V_a is also a storage function. V_a solves (14) with equality.

Under additional assumptions, stability results can be obtained for dissipative systems [27], [10]. We say that Σ^u is (zero state) *detectable* if $w \equiv 0$ and $\lim_{k \rightarrow \infty} z_k = 0$ implies $\lim_{k \rightarrow \infty} x_k = 0$. Σ^u is *asymptotically stable* if $w \equiv 0$, implies $\lim_{k \rightarrow \infty} x_k = 0$ for any initial condition.

Theorem 3.7 *Let $u \in \mathcal{S}$. If Σ^u is finite gain dissipative and detectable, then Σ^u is asymptotically stable.*

PROOF. Setting $w \equiv 0$ in (15) and using the non-negativity of V we get

$$\sum_{i=0}^{k-1} |z_{i+1}|^2 \leq V(x_0), \text{ for all } k > 0,$$

for any initial condition x_0 . This implies $\{z_k\} \in \ell_2((0, \infty), \mathbf{R}^q)$, and so $\lim_{k \rightarrow \infty} z_k = 0$. By detectability, we obtain $\lim_{k \rightarrow \infty} x_k = 0$. \square

Remark 3.8 In general, detectability (or observability) is a key property required for asymptotic stability as it is related to the positive definiteness of the storage functions [10], [24], but is difficult to check, and will depend on the controller $u \in \mathcal{S}$. Detectability holds trivially in the uniformly coercive case: $-\nu_0 + \nu_1 |x|^2 \leq |l(x, u, w)|^2$, where $\nu_0 \geq 0$, $\nu_1 > 0$.

3.3.2 Solution to the State Feedback Robust Control Problem

It is clear from the previous section that the state feedback robust control problem can be solved provided a stabilizing feedback controller can be found which renders the closed loop system finite gain dissipative. The next two theorems give both necessary and sufficient conditions in terms of a controlled version of the dissipation inequality, under a suitable detectability condition.

Theorem 3.9 (Necessity) *If a controller $u^s \in \mathcal{S}$ solves the state feedback robust control problem then there exists a function $V(x)$ such that $V(x) \geq 0$, $V(0) = 0$, and*

$$(17) \quad V(x) \geq \inf_{u \in U} \sup_{w \in \mathbb{R}^r} \left\{ V(b(x, u, w)) - \gamma^2 |w|^2 + |l(x, u, w)|^2 \right\}.$$

PROOF. Since Σ^{u^s} is finite gain, the Bounded Real Lemma 3.4 implies the existence of a storage function V_a satisfying the dissipation inequality (14):

$$V_a(x) \geq \sup_{w \in \mathbb{R}^r} \left\{ V_a(b(x, \bar{u}^s(x), w)) + |l(x, \bar{u}^s(x), w)|^2 - \gamma^2 |w|^2 \right\}.$$

Therefore, V_a satisfies (17). Also, it is clear that $V_a(0) = 0$ and $V_a(x) \geq 0$. \square

Theorem 3.10 (Sufficiency) *Assume that V is a solution of (17) satisfying $V(x) \geq 0$ and $V(0) = 0$. Let $\bar{u}^*(x)$ be a control value which achieves the minimum in (17). Then the controller $u^* \in \mathcal{S}$ defined by $\bar{u}^*(x)$ solves the state feedback robust control problem if the closed loop system Σ^{u^*} is detectable.*

PROOF. The closed loop system Σ^{u^*} is finite gain dissipative, since (17) implies (14) for the controller u^* ; that is, the function V satisfies

$$V(x) \geq \sup_{w \in \mathbb{R}^r} \left\{ V(b(x, \bar{u}^*(x), w)) + |l(x, \bar{u}^*(x), w)|^2 - \gamma^2 |w|^2 \right\}.$$

Hence by Theorem 3.4, Σ^{u^*} is finite gain. Theorem 3.7 then shows that Σ^{u^*} is asymptotically stable. Hence u^* solves the state feedback robust control problem. \square

Remark 3.11 The utility of this result is that the controlled dissipation inequality (17) provides (in principle) a recipe for solving the state feedback robust control problem.

4 The Output Feedback Case

We return now to the output feedback robust control problem. As in the state feedback case, we start with a finite time version.

4.1 Finite Time Case

Let $\mathcal{O}_{k,l}$ denote the set of output feedback controllers defined on the time interval $[k, l]$, so $u \in \mathcal{O}_{k,l}$ means that for each $j \in [k, l]$ there exists a function $\bar{u}_j : \mathbf{R}^{(j-k+1)p} \rightarrow U$ such that $u_j = \bar{u}_j(y_{k+1,j})$. For $u \in \mathcal{O}_{0,M-1}$, Σ^u denotes the closed loop system (1).

The *finite time output feedback robust control problem* is: given $\gamma > 0$ and a finite time interval $[0, k]$, find a controller $u \in \mathcal{O}_{0,k-1}$ such that the resulting closed loop system Σ^u achieves the following goal;

Σ^u is *finite gain*, i.e., for each initial condition $x_0 \in \mathbf{R}^n$ the corresponding input-output map $\Sigma_{x_0}^u$ relating w to z is finite gain, which means that there exists a finite quantity $\beta_k^u(x_0)$, with $\beta_k^u(0) = 0$, such that

$$(18) \quad \begin{cases} \sum_{i=0}^{k-1} |z_{i+1}|^2 \leq \gamma^2 \sum_{i=0}^{k-1} |w_i|^2 + \beta_k^u(x_0) \\ \text{for all } w \in \ell_2([0, k-1], \mathbf{R}^r), \end{cases}$$

4.1.1 Dynamic Game

Our aim in this section is to express the output feedback robust control problem in terms of a dynamic game.

We introduce the function space

$$\mathcal{E} = \{p : \mathbf{R}^n \rightarrow \mathbf{R}^*\},$$

and define for each $x \in \mathbf{R}^n$ a function $\delta_x \in \mathcal{E}$ by

$$\delta_x(\xi) \triangleq \begin{cases} 0 & \text{if } \xi = x, \\ -\infty & \text{if } \xi \neq x. \end{cases}$$

For $u \in \mathcal{O}_{0,k-1}$ and $p \in \mathcal{E}$ define the functional $J_{p,k}(u)$ for the system (1) by

$$(19) \quad J_{p,k}(u) \triangleq \sup_{w \in \ell_2([0, k-1], \mathbf{R}^r)} \sup_{x_0 \in \mathbf{R}^n} \left\{ p(x_0) + \sum_{i=0}^{k-1} |z_{i+1}|^2 - \gamma^2 |w_i|^2 \right\}.$$

Remark 4.1 The quantity $p \in \mathcal{E}$ in (19) can be chosen in a way which reflects knowledge of any *a priori* information concerning the initial state x_0 of Σ^u . \square

The finite gain property of Σ^u can be expressed in terms of J as follows.

Lemma 4.2 $\Sigma_{x_0}^u$ is finite gain on $[0, k]$ if and only if there exists a finite quantity $\beta_k^u(x_0)$ such that

$$(20) \quad J_{\delta_{x_0}, k}(u) \leq \beta_k^u(x_0),$$

and $\beta_k^u(0) = 0$.

It is of interest to know when $J_{p,k}(u)$ is finite. For a finite gain system Σ^u , we write

$$\text{dom } J_{p,k}(u) = \{p \in \mathcal{E} : (p, 0), (p, \beta_k^u) \text{ finite}\},$$

where we use the pairing [13]

$$(21) \quad (p, q) = \sup_{x \in \mathbb{R}^n} \{p(x) + q(x)\}.$$

Lemma 4.3 If each map $\Sigma_{x_0}^u$ is finite gain on $[0, k]$, then

$$(22) \quad (p, 0) \leq J_{p,k}(u) \leq (p, \beta_k^u),$$

and so $J_{p,k}(u)$ is finite for $p \in \text{dom } J_{p,k}(u)$.

PROOF. Set $w \equiv 0$ in (19) to deduce $(p, 0) \leq J_{p,k}(u)$. Next, select $w \in \ell_2([0, k-1], \mathbb{R}^r)$ and $x_0 \in \mathbb{R}^n$. Then (18) implies

$$p(x_0) + \sum_{i=0}^{k-1} |z_{i+1}|^2 - \gamma^2 |w_i|^2 \leq p(x_0) + \beta_k^u(x_0) \leq (p, \beta_k^u).$$

This proves (22). □

The *finite time output feedback dynamic game* is to find a control policy $u \in \mathcal{O}_{0,k-1}$ which minimizes each functional $J_{\delta_{x_0}, k}$. The idea then is that a solution to this game problem will solve the output feedback robust control problem.

4.1.2 Information State Formulation

To solve the game problem, we borrow an idea from stochastic control theory (see, e.g., [5], [15]) and replace the original problem with a new one expressed in terms of a new state variable, viz., an information state [13]; c.f. also [4], [25], [26].

For fixed $y_{1,j} \in \ell_2([1, j], \mathbb{R}^p)$ we define the *information state* $p_j \in \mathcal{E}$ by

$$(23) \quad \begin{aligned} p_j(x) &\triangleq \sup_{w \in \ell_2([0, j-1], \mathbb{R}^r)} \sup_{x_0 \in \mathbb{R}^n} \left\{ p_0(x_0) + \sum_{i=0}^{j-1} |z_{i+1}|^2 - \gamma^2 |w_i|^2 \right. \\ &\quad \left. : x_j = x, h(x_i, u_i, w_i) = y_{i+1}, 0 \leq i \leq j-1 \right\}. \end{aligned}$$

If Σ^u is finite gain, then

$$-\infty \leq p_j(x) \leq (p_0, \beta_k^u) < +\infty.$$

A finite lower bound depends on possible degeneracies in the system (1).

In order to write the dynamical equation for p_j , we define $F(p, u, y) \in \mathcal{E}$ by

$$(24) \quad F(p, u, y)(x) = \sup_{\xi \in \mathbb{R}^n} \{p(\xi) + B(\xi, x, u, y)\},$$

where the extended real valued function B is defined by

$$(25) \quad B(\xi, x, u, y) = \sup_{w \in \mathbb{R}^r} \left\{ |l(\xi, u, w)|^2 - \gamma^2 |w|^2 : b(\xi, u, w) = x, h(\xi, u, w) = y \right\}.$$

Here, we use the convention that the supremum over an empty set equals $-\infty$.

Lemma 4.4 *The information state is the solution of the following recursion:*

$$(26) \quad \begin{cases} p_j = F(p_{j-1}, u_{j-1}, y_j), & j \in [1, k], \\ p_0 \in \mathcal{E}. \end{cases}$$

PROOF. The result is proven by induction. Assume the assertion is true for $0, \dots, j-1$; we must show that p_j defined by (23) equals $F(p_{j-1}, u_{j-1}, y_j)$ defined by (24). Now

$$\begin{aligned} F(p_{j-1}, u_{j-1}, y_j)(x) &= \sup_{\xi \in \mathbb{R}^n} \{p_{j-1}(\xi) + B(\xi, x, u_{j-1}, y_j)\} \\ &= \sup_{\xi \in \mathbb{R}^n} \left\{ p_{j-1}(\xi) + \sup_{w_{j-1} \in \mathbb{R}^r} (|l(\xi, u_{j-1}, y_j)|^2 - \gamma^2 |w_{j-1}|^2 : \right. \\ &\quad \left. b(\xi, u_{j-1}, w_{j-1}) = x, h(\xi, u_{j-1}, w_{j-1}) = y_j) \right\} \\ &= p_j(x) \end{aligned}$$

using the definition (23) for p_{j-1} and p_j . □

Remark 4.5 Note that we can write

$$(27) \quad p_j(x) \triangleq \sup_{\xi \in \ell_2([0, j], \mathbb{R}^n)} \left\{ p_0(\xi_0) + \sum_{i=0}^{j-1} B(\xi_i, \xi_{i+1}, u_i, y_{i+1}) : \xi_j = x \right\}.$$

□

We now state the following representation result:

Theorem 4.6 For $u \in \mathcal{O}_{0,j-1}$, $p \in \mathcal{E}$, such that $J_{p,j}(u)$ is finite, we have the representation

$$(28) \quad J_{p,j}(u) = \sup_{y_{1,j} \in \ell_2([1,j], \mathbf{R}^p)} \{(p_j, 0) : p_0 = p\}, \quad j \in [0, k].$$

PROOF. We have

$$\begin{aligned} & \sup_{y_{1,j} \in \ell_2([1,j], \mathbf{R}^p)} \{(p_j, 0) : p_0 = p\} \\ &= \sup_{y_{1,j} \in \ell_2([1,j], \mathbf{R}^p)} \sup_{\xi \in \ell_2([0,j], \mathbf{R}^n)} \left\{ p(\xi_0) + \sum_{i=0}^{j-1} B(\xi_i, \xi_{i+1}, u_i, y_{i+1}) \right\} \\ &= \sup_{w \in \ell_2([0,j-1], \mathbf{R}^r)} \sup_{x_0 \in \mathbf{R}^n} \left\{ p(x_0) + \sum_{i=0}^{j-1} |z_{i+1}|^2 - \gamma^2 |w_i|^2 \right\} \\ &= J_{p,j}(u). \end{aligned}$$

□

Remark 4.7 This representation theorem is a *separation principle*, and is similar to those employed in stochastic control theory, see [15], and in particular, [5], [13]. □

Theorem 4.6 enables us to express the finite gain property of Σ^u in terms of the information state p , as the following corollary shows:

Corollary 4.8 For any output feedback controller $u \in \mathcal{O}_{0,k-1}$, the closed loop system Σ^u is finite gain on $[0, k]$ if and only if the information state p_j satisfies

$$(29) \quad \sup_{y_{1,j} \in \ell_2([1,j], \mathbf{R}^p)} \{(p_j, 0) : p_0 = \delta_{x_0}\} \leq \beta_k^u(x_0), \quad \text{for all } j \in [0, k],$$

for some finite $\beta^u(x_0, k)$ with $\beta_k^u(0) = 0$.

Remark 4.9 In view of the above, the name “information state” for p is justified. Indeed, p_j contains all the information relevant to the key finite gain property of Σ^u that is available in the observations $y_{1,j}$. □

Remark 4.10 We now regard the information state dynamics (26) as a new (infinite dimensional) control system Ξ , with control u and disturbance y . The state p_j and disturbance y_j are available to the controller, so the original output feedback dynamic game is equivalent to a new one with *full information*. The cost is now the RHS of (28). The analogue in stochastic control theory is the dynamical equation for the conditional density (or variant), and y becomes white noise under a reference probability measure [13], [15]. □

Now that we have introduced the new state variable p , we need an appropriate class $\mathcal{I}_{i,l}$ of controllers which feedback this new state variable. A control u belongs to $\mathcal{I}_{i,l}$ if for each $j \in [i, l]$ there exists a map \bar{u}_j from a subset of \mathcal{E}^{j-i+1} (sequences $p_{i,j} = p_i, p_{i+1}, \dots, p_j$) into U such that $u_j = \bar{u}_j(p_{i,j})$. Note that since p_j depends only on the observable information $y_{1,j}$, $\mathcal{I}_{0,j-1} \subset \mathcal{O}_{0,j-1}$.

4.1.3 Solution to the Finite Time Output Feedback Robust Control Problem

In this subsection we use dynamic programming to obtain necessary and sufficient conditions for the solution of the output feedback robust control problem. We make use of the dynamic programming approach used in [13] to solve the output feedback dynamic game problem. The value function is given by

$$(30) \quad W_j(p) = \inf_{u \in \mathcal{O}_{0,j-1}} \sup_{y \in \ell_2([1,j], \mathbb{R}^p)} \{ (p_j, 0) : p_0 = p \},$$

for $j \in [0, k]$, and the corresponding dynamic programming equation is

$$(31) \quad \begin{cases} W_j(p) = \inf_{u \in U} \sup_{y \in \mathbb{R}^p} \{ W_{j-1}(F(p, u, y)) \}, & j \in [1, k], \\ W_0(p) = (p, 0). \end{cases}$$

For a function $W : \mathcal{E} \rightarrow \mathbb{R}^*$, we write

$$\text{dom } W = \{ p \in \mathcal{E} : W(p) \text{ finite} \}.$$

Theorem 4.11 (Necessity) *Assume that $u^\circ \in \mathcal{O}_{0,k-1}$ solves the finite time output feedback robust control problem. Then there exists a solution W to the dynamic programming equation (31) such that $\text{dom } J_{p,k}(u^\circ) \subset \text{dom } W_j$, $W_j(\delta_0) = 0$, $W_j(p) \geq (p, 0)$, $j \in [0, k]$.*

PROOF. For $p \in \text{dom } J_{p,k}(u^\circ)$, define $W_j(p)$ by (30), i.e.,

$$W_j(p) = \inf_{u \in \mathcal{O}_{0,j-1}} J_{p,j}(u).$$

Note the alternative expression for $W_j(p)$:

$$(32) \quad W_j(p) = \inf_{u \in \mathcal{O}_{0,j-1}} \sup_{w \in \ell_2([0,j-1], \mathbb{R}^r)} \sup_{x_0 \in \mathbb{R}^n} \left\{ p(x_0) + \sum_{i=0}^{j-1} |z_{i+1}|^2 - \gamma^2 |w_i|^2 \right\}.$$

For $u = u^\circ$ we see that, using the finite gain property for Σ^{u° ,

$$\begin{aligned} W_j(p) &\leq \sup_{w \in \ell_2([0,j-1], \mathbb{R}^r)} \sup_{x_0 \in \mathbb{R}^n} \left\{ p(x_0) + \sum_{i=0}^{j-1} |z_{i+1}|^2 - \gamma^2 |w_i|^2 \right\} \\ &\leq (p, \beta_k^{u^\circ}). \end{aligned}$$

Thus $\text{dom } J_{p,k}(u^\circ) \subset \text{dom } W_j$. Also, we have

$$W_j(p) \geq (p, 0).$$

Since $\beta_k^{u^\circ}(0) = 0$, $(\delta_0, 0) = 0$, we have $W_j(\delta_0) = 0$. Finally, the proof of Theorem 4.4, [13] shows that W_j is the unique solution of the dynamic programming equation (31). \square

Theorem 4.12 (Sufficiency) *Assume there exists a solution W to the dynamic programming equation (31) such that $\delta_x \in \text{dom } W_j$ for all $x \in \mathbf{R}^n$, $W_j(\delta_0) = 0$, $W_j(p) \geq (p, 0)$, $j \in [0, k]$. Let $u^* \in \mathcal{I}_{0,k-1}$ be a policy such that $u_j^* = \bar{u}_{k-j}^*(p_j)$, where $\bar{u}_j^*(p)$ achieves the minimum in (31); $j = 0, \dots, k-1$. Then u^* solves the finite time output feedback robust control problem.*

PROOF. Following the proof of Theorem 4.6 of [13], we see that

$$W_k(p) = J_{p,k}(u^*) \leq J_{p,k}(u)$$

for all $u \in \mathcal{O}_{0,k-1}$, $p \in \text{dom } W_k$. Now

$$\sup_{y \in \ell_2([1,k], \mathbf{R}^p)} \{ (p_k, 0) : p_0 = \delta_{x_0}, u = u^* \} \leq W_k(\delta_{x_0}),$$

which implies by Corollary 4.8 that Σ^{u^*} is finite gain with $\beta_k^{u^*}(x_0) = W_k(\delta_{x_0})$, and hence u^* solves the finite time output feedback robust control problem. \square

Remark 4.13 Note that the controller obtained in Theorem 4.12 is an *information state feedback* controller. \square

Corollary 4.14 *If the finite time output feedback robust control problem is solvable by an output feedback controller $u^\circ \in \mathcal{O}_{0,k-1}$, then it is also solvable by an information state feedback controller $u^* \in \mathcal{I}_{0,k-1}$.*

4.2 Infinite Time Case

Again, we would like to solve the infinite time problem by passing to the limit

$$(33) \quad \lim_{k \rightarrow \infty} W_k(p) = W(p),$$

where $W_k(p)$ is defined by (30), to obtain a stationary version of the dynamic programming equation (31), viz.,

$$(34) \quad W(p) = \inf_{u \in U} \sup_{y \in \mathbf{R}^p} \{ W(F(p, u, y)) \}.$$

However, for technical reasons, this is not quite what we do. Instead, we will minimize the functional

$$(35) \quad J_p(u) = \sup_{k \geq 0} J_{p,k}(u)$$

over $u \in \mathcal{O}$. Here, \mathcal{O} denotes output feedback controllers u such that for each k , $u_k = \bar{u}_k(y_{1,k})$ for some map \bar{u}_k from \mathbf{R}^{pk} into U . This makes sense in view of the following lemma, whose proof is an easy consequence of the definitions (c.f. Corollary 4.8).

Lemma 4.15 *For any output feedback controller $u \in \mathcal{O}$, the closed loop system Σ^u is finite gain if and only if the information state p_k satisfies*

$$(36) \quad \sup_{k \geq 0} \sup_{y_{1,k} \in \ell_2([1,k], \mathbf{R}^p)} \{(p_k, 0) : p_0 = \delta_{x_0}\} \leq \beta^u(x_0),$$

for some finite $\beta^u(x_0)$ with $\beta^u(0) = 0$.

Our results will be expressed in terms of an appropriate dissipation inequality, and so in the next section we formulate an appropriate version of the Bounded Real Lemma for the information state system.

4.2.1 Bounded Real Lemma

Let \mathcal{I} denote the class of information state feedback controllers u such that $u_k = \bar{u}(p_k)$, for some function \bar{u} from a subset of \mathcal{E} into U .

From Lemma 4.15, we say that the information state system Ξ^u ((26) with information state feedback $u \in \mathcal{I}$) is *finite gain* if and only if the information state p_k satisfies (36) for some finite $\beta^u(x_0)$ with $\beta^u(0) = 0$. For a finite gain system Σ^u , we write

$$\text{dom } J_p(u) = \{p \in \mathcal{E} : (p, 0), (p, \beta^u) \text{ finite}\},$$

We say that the information state system Ξ^u is *finite gain dissipative* if there exists a function (called a *storage function*) $W(p)$ such that $\text{dom } W$ contains δ_x for all $x \in \mathbf{R}^n$, $W(p) \geq (p, 0)$, $W(\delta_0) = 0$, and satisfies the dissipation inequality

$$(37) \quad W(p) \geq \sup_{y \in \mathbf{R}^p} \{W(F(p, \bar{u}(p), y))\}.$$

Note that if Ξ^u is finite gain dissipative and $p \in \text{dom } W$, then $F(p, \bar{u}(p), y) \in \text{dom } W$ for all $y \in \mathbf{R}^p$. Consequently, $p_0 \in \text{dom } W$ implies $p_k \in \text{dom } W$ for all $k > 0$.

Theorem 4.16 (Bounded Real Lemma.) *Let $u \in \mathcal{I}$. Then the information state system Ξ^u is finite gain dissipative if and only if it is finite gain.*

PROOF. Assume that Ξ^u is finite gain dissipative. Then (37) implies

$$(38) \quad W(p_k) \leq W(p_0)$$

for all $k > 0$ and all $y \in \ell_2([1, k], \mathbf{R}^p)$. Setting $p_0 = \delta_{x_0}$ and using the inequality $W(p) \geq (p, 0)$ we get

$$(p_k, 0) \leq W(\delta_{x_0})$$

for all $k > 0$, $y \in \ell_2([1, k], \mathbf{R}^p)$. Therefore Ξ^u is finite gain, with $\beta^u(x_0) \triangleq W(\delta_{x_0})$.

Conversely, assume that Ξ^u is finite gain. Then

$$(p, 0) \leq J_{p,k}(u) \leq (p, \beta^u)$$

for all $k \geq 0$, $p \in \text{dom } J_p(u)$. Write $W_k(p) = J_{p,k}(u)$, so that

$$(p, 0) \leq W_k(p) \leq (p, \beta^u), \quad k \geq 0, \quad p \in \text{dom } J_p(u).$$

Now W_k is monotone non-decreasing:

$$W_{k-1}(p) \leq W_k(p).$$

To see this, note that

$$W_k(p) = \sup_{w \in \ell_2([0, k-1], \mathbf{R}^r)} \sup_{x_0 \in \mathbf{R}^n} \left\{ p(x_0) + \sum_{i=0}^{k-1} |z_{i+1}|^2 - \gamma^2 |w_i|^2 \right\}.$$

Then given $\varepsilon > 0$, choose $w' \in \ell_2([0, k-2], \mathbf{R}^r)$ and x'_0 such that

$$W_{k-1}(p) \leq p(x_0) + \sum_{i=0}^{k-2} |z'_{i+1}|^2 - \gamma^2 |w'_i|^2 + \varepsilon,$$

and define $w \in \ell_2([0, k-1], \mathbf{R}^r)$ by setting $w = w'$ on $[0, k-2]$ and $w_{k-1} = 0$, and let $x_0 = x'_0$. Then

$$\begin{aligned} W_k(p) &\geq p(x'_0) + \sum_{i=0}^{k-1} |z_{i+1}|^2 - \gamma^2 |w_i|^2 \\ &\geq p(x_0) + \sum_{i=0}^{k-2} |z'_{i+1}|^2 - \gamma^2 |w'_i|^2 + |z_k|^2 \\ &\geq W_{k-1}(p) - \varepsilon. \end{aligned}$$

Since $\varepsilon > 0$ is arbitrary, the monotonicity assertion is verified.

Therefore the limit

$$(39) \quad W_a(p) = \lim_{k \rightarrow \infty} W_k(p)$$

exists and is finite on $\text{dom } W_a$, which contains $\text{dom } J_p(u)$. We now show that W_a satisfies (37). Fix $p \in \text{dom } W_a$, $y \in \mathbf{R}^p$, $\varepsilon > 0$. Select $k \geq 0$ and $\tilde{y}_{1,k}$ such that

$$W_a(F(p, \bar{u}(p), y)) \leq (\tilde{p}_{k-1}, 0) + \varepsilon,$$

where \tilde{p}_j , $j = 0, 1, \dots, k-1$ is the corresponding information state sequence with $\tilde{p}_0 = F(p, \bar{u}(p), y)$. Define $y_{1,k}$ by setting

$$y_i = \begin{cases} y & \text{if } i = 1, \\ \tilde{y}_{i-1} & \text{if } i \geq 2, \end{cases}$$

and let p_j , $j = 0, 1, \dots, k$ denote the corresponding information state trajectory with $p_0 = p$. Then

$$\begin{aligned} W_a(p) &\geq (p_k, 0) \\ &= (\tilde{p}_{k-1}, 0) \\ &\geq W_a(F(p, \bar{u}(p), y)) - \varepsilon. \end{aligned}$$

Since y is arbitrary, we have

$$W_a(p) \geq \sup_{y \in \mathbf{R}^p} W_a(F(p, \bar{u}(p), y)) - \varepsilon.$$

This inequality implies that W_a solves (37). (Actually, W_a solves (37) with equality.) By definition, $W_a(p) \geq (p, 0)$. This and (36) imply $W_a(\delta_0) = 0$. Thus Ξ^u is finite gain dissipative. \square

Remark 4.17 The function W_a defined by (39) is called the *available storage* for the information state system. If Ξ^u is finite gain dissipative with storage function W , then $W_a \leq W$, and W_a is also a storage function. W_a solves (37) with equality.

As in the case of complete state information, we can deduce stability results for the closed loop system Σ^u . Stability here means *internal stability*, and so we must concern ourselves with the stability of the information state system as well.

For the remainder of the paper, we will assume that h satisfies the linear growth condition:

$$(40) \quad |h(x, u, w)| \leq C(|x| + |w|).$$

We say that Σ^u is (zero state) *z-detectable* (resp. *ℓ_2 -z-detectable*) if $w \equiv 0$ and $\lim_{k \rightarrow \infty} z_k = 0$ implies $\lim_{k \rightarrow \infty} x_k = 0$ (resp. $\{z_k\} \in \ell_2([0, \infty), \mathbf{R}^q)$ implies $\{x_k\} \in \ell_2([0, \infty), \mathbf{R}^n)$) and asymptotically stable if $w \equiv 0$ implies $\lim_{k \rightarrow \infty} x_k = 0$ for any initial condition.

For $u \in \mathcal{O}$ and $y \in \ell_2([0, \infty), \mathbf{R}^p)$, Σ^u is *uniformly (w, y) -reachable* if for all $x \in \mathbf{R}^n$ there exists $0 \leq \alpha(x) < +\infty$ such that for all $k \geq 0$ sufficiently large there exists $x_0 \in \mathbf{R}^n$ and $w \in \ell_2([0, k-1], \mathbf{R}^r)$ such that $x(0) = x_0$, $x(k) = x$, $h(x_i, u_i, w_i) = y_{i+1}$, $i = 0, \dots, k-1$, and

$$(41) \quad |x_0|^2 + \sum_{i=0}^{k-1} |w_i|^2 \leq \alpha(x).$$

Given inputs $u \in \mathcal{O}$ and $y \in \ell_2([0, \infty), \mathbf{R}^p)$, we say that the information state system Ξ^u is *stable* if for each $x \in \mathbf{R}^n$ there exists $K_x \geq 0$, $C_x \geq 0$ such that

$$(42) \quad |p_k(x)| \leq C_x \text{ for all } k \geq K_x,$$

provided the initial value p_0 satisfies the growth conditions

$$(43) \quad -a'_1|x|^2 - a'_2 \leq p_0(x) \leq -a_1|x|^2 + a_2,$$

where $a_1, a'_1, a_2, a'_2 \geq 0$.

Theorem 4.18 *Let $u \in \mathcal{I}$. If Ξ^u is finite gain dissipative and Σ^u is z -detectable, then Σ^u is asymptotically stable. If Ξ^u is finite gain dissipative and Σ^u is ℓ_2 - z -detectable and uniformly (w, y) -reachable, then Ξ^u is stable.*

PROOF. Inequality (38) implies

$$(44) \quad \sup_{w \in \ell_2([0, k-1], \mathbf{R}^r)} \sup_{x_0 \in \mathbf{R}^n} \left\{ p(x_0) + \sum_{i=0}^{k-1} |z_{i+1}|^2 - \gamma^2 |w_i|^2 \right\} \leq W(p),$$

for all $k > 0$. Let $x_0 \in \mathbf{R}^n$ and select $p = \delta_{x_0}$. Then (44) gives, with $w \equiv 0$,

$$\sum_{i=0}^{k-1} |z_{i+1}|^2 \leq W(\delta_{x_0}) < +\infty,$$

for all $k > 0$. This implies $\{z_k\} \in \ell_2((0, \infty), \mathbf{R}^q)$, and so $\lim_{k \rightarrow \infty} z_k = 0$. By z -detectability, we obtain $\lim_{k \rightarrow \infty} x_k = 0$. Therefore Σ^u is asymptotically stable.

Also, ℓ_2 - z -detectability implies $\{x_k\} \in \ell_2([0, \infty), \mathbf{R}^n)$, and by assumption (40), $\{y_k\} \in \ell_2([1, \infty), \mathbf{R}^p)$ since $w \equiv 0$.

So now suppose that $\{y_k\} \in \ell_2([0, \infty), \mathbf{R}^p)$. We wish to show that Ξ^u is stable. The dissipation inequality implies

$$p_k(x) \leq (p_k, 0) \leq W(p) < +\infty$$

for all $p_0 \in \text{dom } W$, $k \geq 0$. For the lower bound, the hypothesis imply, given x , for all k sufficiently large there exists x_0 and w such that $x(0) = x_0$, $x(k) = x$, and

$$|x_0|^2 + \sum_{i=0}^{k-1} |w_i|^2 \leq \alpha(x)$$

for some finite non-negative α . Thus

$$\begin{aligned} p_k(x) &\geq p_0(x_0) - \gamma^2 \sum_{i=0}^{k-1} |w_i|^2 \\ &\geq -(a'_1 + \gamma^2)\alpha(x) - a'_2 \end{aligned}$$

for all k sufficiently large. Therefore Ξ^u is stable. \square

Remark 4.19 The behavior we are attempting to capture here is that of eventual finiteness (and in fact boundedness) of the information state. The criteria used to imply stability are modelled on those used in the state feedback case, and are of course difficult to check in practice. These conditions simplify greatly under appropriate nondegeneracy assumptions. Note that it is feasible that Σ^u is stable, with Ξ^u unstable; this corresponds to an unstable stabilizing controller.

4.2.2 Solution to the Output Feedback Robust Control Problem

We begin this subsection with a proposition which asserts that if the output feedback robust control problem is solvable by an information state feedback controller, then there exists a solution to the dissipation inequality (45) below, using the Bounded Real Lemma 4.16. However, this result is not adequate for a necessity theorem, since it is expressed *a priori* in terms of an information state feedback controller. The necessity theorem (Theorem 4.21 below) asserts the existence of a solution of the dissipation inequality assuming only that the output feedback robust control problem is solved by an output feedback controller, which need not necessarily be an information state feedback controller.

Proposition 4.20 *If a controller $u^i \in \mathcal{I}$ solves the output feedback robust control problem, then there exists a function $W(p)$ such that $\text{dom } W$ contains δ_x for all $x \in \mathbf{R}^n$, $W(p) \geq (p, 0)$, $W(\delta_0) = 0$, and*

$$(45) \quad W(p) \geq \inf_{u \in U} \sup_{y \in \mathbf{R}^p} \{ W(F(p, u, y)) \}.$$

PROOF. The Bounded Real Lemma 4.16 implies the existence of a storage function W_a satisfying the dissipation inequality (37):

$$W_a(p) \geq \sup_{y \in \mathbf{R}^p} \{ W_a(F(p, \bar{u}^i(p), y)) \}.$$

Therefore W_a satisfies (45). Also, we have $\delta_x \in \text{dom } W_a$ for all $x \in \mathbf{R}^n$, $W_a(p) \geq (p, 0)$, and $W_a(\delta_0) = 0$. \square

Theorem 4.21 (Necessity) *Assume that there exists a controller $u^o \in \mathcal{O}$ which solves the output feedback robust control problem. Then there exists a function $W(p)$ which is finite on $\text{dom } J_p(u^o)$, satisfies $W(p) \geq (p, 0)$, $W(\delta_0) = 0$, and solves the dissipation inequality (45).*

PROOF. Define

$$W(p) = \inf_{u \in \mathcal{O}} J_p(u),$$

where $J_p(u)$ is defined by (35). Then we have

$$(p, 0) \leq W(p) \leq J_p(u^o) \leq (p, \beta^{u^o}),$$

and so W is finite on $\text{dom } J_p(u^o)$. Clearly $W(p) \geq (p, 0)$ and $W(\delta_0) = 0$. It remains to show that W satisfies (45).

Let $\varepsilon > 0$. Choose $u \in \mathcal{O}$ such that

$$(46) \quad W(p) \geq \sup_{k \geq 0, y \in \ell_2([1, k], \mathbf{R}^r)} \{ (p_k, 0) : p_0 = p \} - \varepsilon.$$

Select any $y \in \mathbf{R}^p$. For any sequence $\tilde{y}_1, \tilde{y}_2, \dots$, define a sequence y_1, y_2, \dots by

$$y_i = \begin{cases} y & \text{if } i = 1, \\ \tilde{y}_{i-1} & \text{if } i \geq 2, \end{cases}$$

and a control $\tilde{u} \in \mathcal{O}$ by

$$\tilde{u}_i(\tilde{y}_1, \dots, \tilde{y}_i) = u_{i+1}(y, \tilde{y}_1, \dots, \tilde{y}_i).$$

Let p_j and \tilde{p}_j denote the information state sequences corresponding to $p_0 = p, u, y_1, y_2, \dots$, and $\tilde{p}_0 = F(p, u_0, y), \tilde{u}, \tilde{y}_1, \tilde{y}_2, \dots$ respectively. Note that $p_k = \tilde{p}_{k-1}$. Now choose $\tilde{y}_1, \tilde{y}_2, \dots$ and $k \geq 1$ such that

$$W(\tilde{p}_0) \leq (\tilde{p}_{k-1}, 0) + \varepsilon.$$

Then

$$(47) \quad (p_k, 0) \geq W(F(p, u_0, y)) - \varepsilon.$$

Combining inequalities (46) and (47), we obtain

$$W(p) \geq W(F(p, u_0, y)) - 2\varepsilon.$$

Since y was selected arbitrarily, we get

$$W(p) \geq \sup_{y \in \mathbf{R}^p} W(F(p, u_0, y)) - 2\varepsilon,$$

and therefore

$$W(p) \geq \inf_{u \in U} \sup_{y \in \mathbf{R}^p} W(F(p, u, y)) - 2\varepsilon.$$

From this, we see that W satisfies (45), since $\varepsilon > 0$ is arbitrary. \square

Theorem 4.22 (Sufficiency) *Assume that W is a solution of (45) satisfying $\delta_x \in \text{dom } W$ for all $x \in \mathbf{R}^n$, $W(p) \geq (p, 0)$ and $W(\delta_0) = 0$. Let $\bar{u}^*(p)$ be a control value which achieves the minimum in (45). Then the controller $u^* \in \mathcal{I}$ defined by $\bar{u}^*(p)$ solves the information state feedback robust control problem if the closed loop system Σ^{u^*} is ℓ_2 - z -detectable and uniformly (w, y) -reachable.*

PROOF. The information state system Ξ^{u^*} is finite gain dissipative, since (45) implies (37) for the controller u^* . Hence by Theorem 4.16, Ξ^{u^*} is finite gain. Theorem 4.18 then shows that Σ^{u^*} is asymptotically stable and Ξ^{u^*} is stable. Hence u^* solves the information state feedback robust control problem. \square

Remark 4.23 As in the state feedback case (§3.3.2), the significance of this result is that the controlled dissipation inequality (45) provides (in principle) a recipe for solving the information state feedback robust control problem.

Corollary 4.24 *If the output feedback robust control problem is solvable by an output-feedback controller $u^o \in \mathcal{O}$, then it is also solvable by an information state feedback controller $u^* \in \mathcal{I}$.*

References

- [1] B.D.O. Anderson and S. Vongpanitlerd, "Network Analysis and Synthesis", Prentice-Hall, Englewood Cliffs, 1973.
- [2] J.A. Ball and J.W. Helton, *Nonlinear H_∞ Control Theory for Stable Plants*, Math. Control Signals Systems, **5** (1992) 233—261.
- [3] J.A. Ball, J.W. Helton and M.L. Walker, *H_∞ Control for Nonlinear Systems with Output Feedback*, preprint, 1991.
- [4] T. Basar and P. Bernhard, " H^∞ -Optimal Control and Related Minimax Design Problems: A Dynamic Game Approach", Birkhauser, Boston, 1991.
- [5] A. Bensoussan and J.H. van Schuppen, *Optimal Control of Partially Observable Stochastic Systems with an Exponential-of-Integral Performance Index*, SIAM J. Control Optim., **23** (1985) 599—613.
- [6] J.C. Doyle, K. Glover, P.P. Khargonekar and B.A. Francis, *State-Space Solutions to Standard H_2 and H_∞ Control Problems*, IEEE Trans. Aut. Control, **34** (8) (1989) 831—847.
- [7] K. Glover and J.C. Doyle, *State Space Formulae for all Stabilizing Controllers that Satisfy an H_∞ Norm Bound and Relations to Risk Sensitivity*, Systems and Control Letters, **11** (1988) 167—172.
- [8] B.A. Francis, "A Course in H_∞ Control Theory", Lecture Notes in Control and Information Science, Vol. 88, Springer Verlag, Berlin, 1987.
- [9] O. Hijab, "Minimum Energy Estimation", Ph.D. Dissertation, Univ. California, Berkeley, 1980.
- [10] D.J. Hill and P.J. Moylan, *The Stability of Nonlinear Dissipative Systems*, IEEE Trans. Automat. Contr., **21** (1976) 708—711.
- [11] D.J. Hill and P.J. Moylan, *Connections Between Finite-Gain and Asymptotic Stability*, IEEE Trans. Automat. Contr., **25** (1980) 931—936.
- [12] A. Isidori and A. Astolfi, *Disturbance Attenuation and H_∞ Control via Measurement Feedback in Nonlinear Systems*, IEEE Trans. Automat. Contr., **37** 9 (1992) 1283—1293.
- [13] M.R. James, J.S. Baras and R.J. Elliott, *Risk-Sensitive Control and Dynamic Games for Partially Observed Discrete-Time Nonlinear Systems*, to appear, IEEE Trans. Automatic Control.
- [14] M.R. James, J.S. Baras and R.J. Elliott, *Output Feedback Risk-Sensitive Control and Differential Games for Continuous-Time Nonlinear Systems*, preprint, 1993.

- [15] P.R. Kumar and P. Varaiya, "Stochastic Systems: Estimation, Identification, and Adaptive Control", Prentice-Hall, Englewood Cliffs, 1986.
- [16] D.J.N. Limebeer, B.D.O. Anderson, P.P. Khargonekar and M. Green, *A Game Theoretic Approach to H_∞ Control for Time Varying Systems*, SIAM J. Control Optim., **30** (1992) 262—283.
- [17] R.E. Mortensen, *Maximum Likelihood Recursive Nonlinear Filtering*, J. Opt. Theory Appl., **2** (1968) 386—394.
- [18] I.R. Petersen, *Disturbance Attenuation and H_∞ Optimization: A Design Method Based on the Algebraic Riccati Equation*, IEEE Trans. Aut. Control, **32** (1987) 427—429.
- [19] I.R. Petersen, B.D.O. Anderson and E.A. Jonckheere, *A First Principles Solution to the Non-Singular H_∞ Control Problem*, Int. Journal Robust Nonlinear Control, **1** (1991) 171—185.
- [20] I. Rhee and J.L. Speyer, *A Game Theoretic Approach to a Finite-Time Disturbance Attenuation Problem*, IEEE Trans. Aut. Control, **36** (1991) 1021—1032.
- [21] A.J. van der Schaft, *On a State Space Approach to Nonlinear H_∞ Control*, Systems and Control Letters, **16** (1991) 1—8.
- [22] A.J. van der Schaft, *L_2 Gain Analysis of Nonlinear Systems and Nonlinear State Feedback H_∞ Control*, IEEE Trans. Aut. Control, **AC-37** (6) (1992) 770—784.
- [23] A.J. van der Schaft, *Nonlinear State Space H_∞ Control Theory*, to appear in: Perspectives in Control, H.L. Trentelman and J.C. Willems Eds., Birkhauser, Series: Progress in Systems and Control.
- [24] M. Vidyasagar, "Nonlinear Systems Analysis", 2nd Edition, Prentice-Hall, Englewood Cliffs, 1993.
- [25] P. Whittle, *Risk-Sensitive Linear/Quadratic/Gaussian Control*, Adv. Appl. Prob., **13** (1981) 764—777.
- [26] P. Whittle, *A Risk-Sensitive Maximum Principle: The Case of Imperfect State Observation*, IEEE Trans. Aut. Control, **36** (1991) 793—801.
- [27] J.C. Willems, *Dissipative Dynamical Systems Part I: General Theory*, Arch. Rational Mech. Anal., **45** (1972) 321—351.
- [28] G. Zames, *Feedback and Optimal Sensitivity: Model Reference Transformations, Multiplicative Seminorms, and Approximate Inverses*, IEEE Trans. Aut. Control, **26** (1981) 301—320.
- [29] G. Zames and B.A. Francis, *Feedback, Minimax Sensitivity, and Optimal Robustness*, IEEE Trans. Aut. Control, **28** (1983) 585—601.